

# Precision, Zero-Drift Instrumentation Amplifier

### **FEATURES**

■ Offset Voltage: 10µV Max

Offset Voltage Drift: 100nV/°C Max

Bias Current: 65pA Max
Offset Current: 65pA Max
Gain Nonlinearity: 20ppm Max
Gain Error: ±0.075% Max

■ CMRR: 90dB

0.1Hz to 10Hz Noise: 1.9µV<sub>P-P</sub>
Single 5V Supply Operation

■ 8-Pin MiniDIP

### **APPLICATIONS**

■ Thermocouple Amplifiers

Strain Gauge Amplifiers

Differential to Single-Ended Converters

### DESCRIPTION

The LTC®1100 is a high precision instrumentation amplifier using zero-drift techniques to achieve outstanding DC performance. The input DC offset is typically  $1\mu V$  while the DC offset drift is typically  $5nV/^{\circ}C$ ; a very low bias current of 65pA is also achieved.

The LTC1100 is self-contained; that is, it achieves a differential gain of 100 without any external gain setting resistor or trim pot. The gain linearity is 20ppm and the gain drift is  $4ppm/^{\circ}C$ . The LTC1100 operates from a single 5V supply up to  $\pm 8V$ . The output typically swings 300mV from its power supply rails with a 10k load.

An optional external capacitor can be added from Pin 7 to Pin 8 to tailor the device's 18kHz bandwidth and to eliminate any unwanted noise pickup.

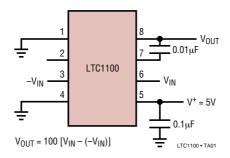
The LTC1100 is also offered in a 16-pin surface mount package with selectable gains of 10 or 100.

The LTC1100 is manufactured using Linear Technology's enhanced LTCMOS<sup>™</sup> silicon gate process.

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## TYPICAL APPLICATION

Single 5V Supply, DC Instrumentation Amplifier

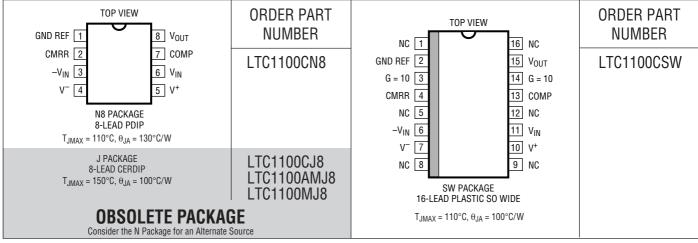


# **ABSOLUTE MAXIMUM RATINGS** (Note 1)

Operating Temperature Range	
LTC1100M/AM (OBSOLETE)	-55°C to 125°C
LTC1100C	-40°C to 85°C
Output Short Circuit Duration	Indefinite

Storage Temperature Ran	ge –65°C to 150°C
Total Supply Voltage (V+	to V <sup>-</sup> ) 18V
Input Voltage	$(V^+ + 0.3V)$ to $(V^ 0.3V)$
Lead Temperature (Solder	ring, 10 sec)300°C

# PACKAGE/ORDER INFORMATION



Consult LTC Marketing for parts specified with wider operating temperature ranges.

# **ELECTRICAL CHARACTERISTICS** The $\bullet$ denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$ . $V_S = \pm 5V$ , $R_L = 10k$ , $C_C = 1000pF$ , unless otherwise noted.

			L	TC1100A0	N	L1			
PARAMETER	CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	UNITS	
Gain Error		•		0.01	0.05 0.10		0.01	0.075 0.150	±% ±%
Gain Nonlinearity		•		3 12	8 30		3 12	20 60	ppm ppm
Input Offset Voltage	(Note 2)			±1	±10		±1	±10	μV
Input Offset Voltage Drift	(Note 2)	•		±5	±100		±5	±100	nV/°C
Input Noise Voltage	DC to 10Hz, T <sub>A</sub> = 25°C			1.9			1.9		μV <sub>P-P</sub>
Input Bias Current		•		2.5	50 120		2.5	65 135	pA pA
Input Offset Current		•		10	50		10	65	pA
Common Mode Rejection Ratio	$V_{CM} = 2.3V \text{ to } -4.7V \text{ (Note 3)}$	•	104	115		90	110		dB
Power Supply Rejection Ratio	$V_S = \pm 2.375 V \text{ to } \pm 8 V$	•	120			105			dB
Output Voltage Swing	$R_L = 2k, V_S = \pm 8V$ $R_L = 10k, V_S = \pm 8V$	•	-7.2 -7.7		6.2 7.5	-7.2 -7.7		6.2 7.5	V
Supply Current		•		2.4 3.4	2.8 4.0		2.4 3.4	3.3 4.5	mA mA
Internal Sampling Frequency				2.8			2.8		kHz
Bandwidth				18			18		kHz

LINEAR TECHNOLOGY

# **ELECTRICAL CHARACTERISTICS** The $\bullet$ denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$ . $V_S = \pm 5V$ , $R_L = 10k$ , $C_C = 1000pF$ , unless otherwise noted.

			LTC1100AMJ (Note 4)			LTC1100MJ			
PARAMETER	CONDITIONS		MIN	TYP `	MAX	MIN	TYP	MAX	UNITS
Gain Error		•		0.01	0.05 0.11		0.01	0.075 0.150	±% ±%
Gain Nonlinearity		•		3	8 40		3	20 65	ppm ppm
Input Offset Voltage	(Note 2)			±1	±10		±1	±10	μV
Input Offset Voltage Drift	(Note 2)	•		±5	±100		±5	±100	nV/°C
Input Noise Voltage	DC to 10Hz, T <sub>A</sub> = 25°C			1.9			1.9		μV <sub>P-P</sub>
Input Bias Current		•		5	50 300		5	65 450	pA pA
Input Offset Current		•			80			120	pA
Common Mode Rejection Ratio	$V_{CM} = -4.7V \text{ to } 2.3V$	•	100			90			dB
Power Supply Rejection Ratio	$V_S = \pm 2.375V \text{ to } \pm 8V$	•	115			95			dB
Output Voltage Swing	$R_L = 10k, V_S = \pm 8V$ $R_L = 2k, V_S = \pm 8V$	•	-7.4 -7.0		7.4 6.0	-7.4 -7.0		7.4 6.0	V
Supply Current		•		2.4	4.2		2.4	3.3 4.6	mA mA
Internal Sampling Frequency				2.8			2.8		kHz
Bandwidth				18			18		kHz

# **ELECTRICAL CHARACTERISTICS** The $\bullet$ denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$ . $V_S = \pm 5V$ , $R_L = 10k$ , $C_C = 1000pF$ , unless otherwise specified.

			L	TC1100A	CS	L			
PARAMETER	CONDITIONS		MIN	TYP	MAX	MIN	TYP	MAX	UNITS
Gain Error	$T_A = 25^{\circ}C, A_V = 100$			0.01	0.05		0.01	0.075	±%
	A <sub>V</sub> =100	•			0.10			0.150	±%
	A <sub>V</sub> =10			0.01	0.04		0.01	0.060	±%
	A <sub>V</sub> =10	•			0.10			0.150	±%
Gain Nonlinearity	$T_A = 25^{\circ}C, A_V = 100$			3	8		3	20	ppm
	A <sub>V</sub> =100	•		12	30		12	60	ppm
	A <sub>V</sub> =10			1	8		1	10	ppm
	A <sub>V</sub> =10	•			25			40	ppm
Input Offset Voltage	(Note 2)			±1	±10		±1	±10	μV
Input Offset Voltage Drift	(Note 2)	•		±5	±100		±5	±100	nV/°C
Input Noise Voltage	DC to 10Hz, T <sub>A</sub> = 25°C			1.9			1.9		μV <sub>P-P</sub>
Input Bias Current				2.5	50		2.5	65	pA
·		•			120			135	pA
Input Offset Current		•		10	50		10	65	pA
Common Mode Rejection Ratio	$V_{CM} = -4.7V$ to 2.3V,								
·	A <sub>V</sub> =100	•	104	115		90	110		dB
	A <sub>V</sub> =10	•	95			85			dB
Power Supply Rejection Ratio	$V_S = \pm 2.375V \text{ to } \pm 8V$	•	120			105			dB

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			L	TC1100A0	S	LTC1100CSW			T	
PARAMETER		CONDITIONS		MIN	TYP	MAX	MIN	TYP	MAX	UNITS
Output Voltage Swing	Swing	$R_L = 10k, V_S = \pm 8V$	•	-7.2		6.2	-7.2		6.2	V
		$R_L = 2k, V_S = \pm 8V$	•	-7.7		7.5	-7.7		7.5	V
Supply Current					2.4	2.8		2.4	3.3	mA
			•		3.4	4.0		3.4	4.5	mA
Internal Sampli	ng Frequency				2.8			2.8		kHz
Bandwidth	G = 100				18			18		kHz
	G = 10				180			180		kHz

**Note 1:** Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

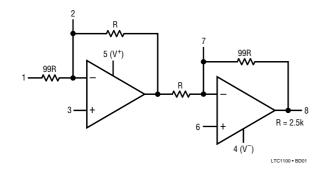
**Note 2:** These parameters are guaranteed by design. Thermocouple effects preclude measurement of these voltage levels in high speed automatic test

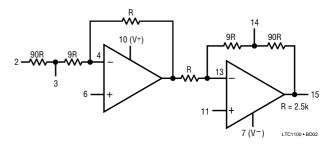
systems.  $\ensuremath{V_{0S}}$  is measured to a limit determined by test equipment capability.

Note 3: See Applications Information, Single Supply Operation.

Note 4: Please consult Linear Technology Marketing.

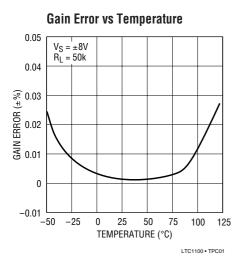
## **BLOCK DIAGRAMS**

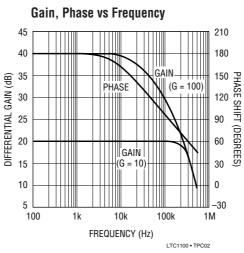


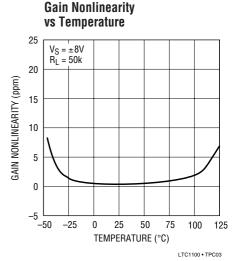


NOTE: FOR A VOLTAGE GAIN OF 10V/V SHORT PIN 2 TO 3, AND PIN 14 TO 15.

# TYPICAL PERFORMANCE CHARACTERISTICS





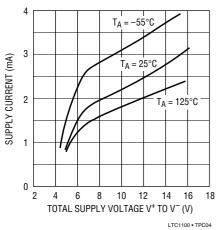


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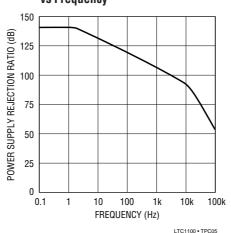
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# TYPICAL PERFORMANCE CHARACTERISTICS

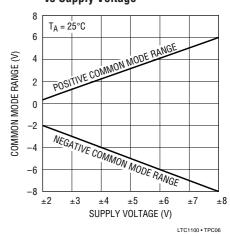
#### Supply Current vs Supply Voltage



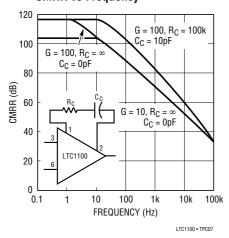
# Power Supply Rejection Ratio vs Frequency



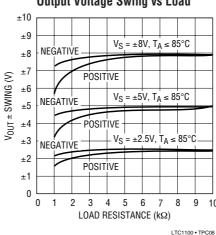
# Common Mode Range vs Supply Voltage



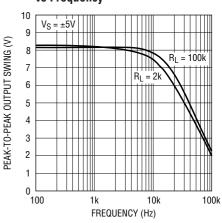
#### **CMRR vs Frequency**





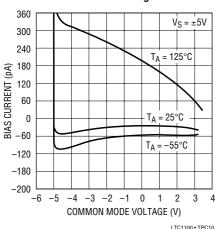


Undistorted Output Swing vs Frequency

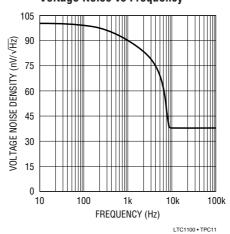


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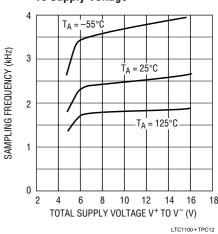
#### Bias Current vs Common Mode Voltage



Voltage Noise vs Frequency



# Internal Sampling Frequency vs Supply Voltage

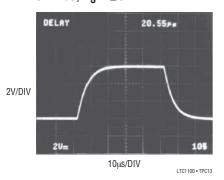


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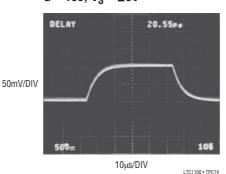


## TYPICAL PERFORMANCE CHARACTERISTICS

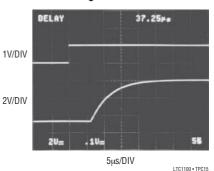
Large-Signal Transient Response  $G = 100, V_S = \pm 5V$ 



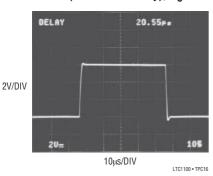
Small-Signal Transient Response  $G = 100, V_S = \pm 5V$ 



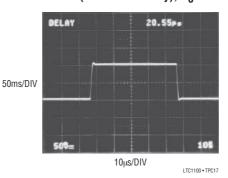
Overload Recovery  $G = 100, V_S = \pm 5V$ 



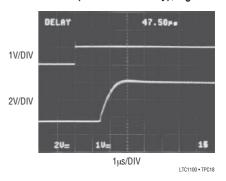
Large-Signal Transient Response  $G = 10 \text{ (LTC1100CS Only)}, V_S = \pm 5V$ 



Small-Signal Transient Response G = 10 (LTC1100CS Only),  $V_S = \pm 5V$ 



Overload Recovery  $G = 10 \text{ (LTC1100CS Only)}, V_S = \pm 5V$ 



## PIN FUNCTIONS

8-Pin DIP (16-Pin SO)

**Pin 1 (2) GND REF:** Connect to system ground. This sets the zero reference for the internal op amps.

**Pin 2 (4) CMRR:** This pin tailors the gain of the internal amplifiers to maximize AC CMRR. For applications which emphasize CMRR requirements, connect a 100k resistor and a 10pF capacitor in series from CMRR to ground. See the Applications section.

Pin 3 (6) -V<sub>IN</sub>: Inverting Input.

Pin 4 (7) V<sup>-</sup>: Negative Supply.

Pin 5 (10) V+: Positive Supply.

Pin 6 (11) V<sub>IN</sub>: Noninverting Input.

**Pin 7 (13) COMP:** This pin reduces the bandwidth of the internal amplifiers for applications at or near DC. Clock feedthrough from the internal sampling clock can also be

suppressed by using the COMP pin. The standard compensation circuit is a capacitor from COMP to  $V_{OUT}$ , sized to provide an RC pole with the internal 247k resistor (22.5k for LTC1100CS in gain-of-10 mode). See the Applications section.

Pin 8 (15) Vour: Signal Output.

#### 16-Pin SO Package Only

(3) G = 10: Short to pin (2) for gain of 10. Leave disconnected for gain of 100.

(14) G = 10: Short to pin (15) for gain of 10. Leave disconnected for gain of 100.

NOTE: *Both* pins must be shorted or open to provide correct gain.

(1),(5),(8),(9),(12),(16) NC: No Internal Connection.

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### APPLICATIONS INFORMATION

#### **Common Mode Rejection**

Due to very precise matching of the internal resistors, no trims are required to obtain a DC CMRR of better than 100dB; however, things change as frequency rises. The inverting amplifier is in a gain of 1.01 (1.1 for gain of 10), while the noninverting amplifier is in a gain of 99 (9 for gain of 10). As frequency rises, the higher gain amplifier hits its gain-bandwidth limit long before the low gain amplifier, degrading CMRR. The solution is straightforward—slow down the inverting amplifier to match the noninverting amp. Figure 1 shows the recommended circuit. The problem is less pronounced in the LTC1100CS in gain-of-10 mode; no CMRR trims are necessary.

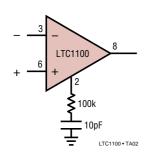


Figure 1. Improving AC CMRR

#### **Overcompensation**

Many instrumentation amplifier applications process DC or low frequency signals only; consequently, the 18kHz (180kHz for G=10) bandwidth of the LTC1100 can be reduced to minimize system errors or reduce transmitted clock noise by using the COMP pin. A feedback cap from COMP to  $V_{OUT}$  will react with the 247k internal resistor (22.5k for G=10) to limit the bandwidth, as in Figure 2.

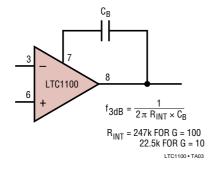


Figure 2. Overcompensation to Reduce System Bandwidth

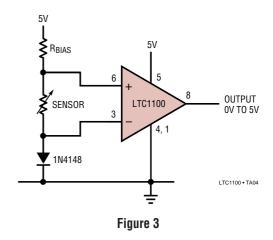
#### **Aliasing**

The LTC1100 is a chopper-stabilized instrumentation amplifier; like all sampled systems it exhibits aliasing behavior for input frequencies at or near the internal sampling frequency. The LTC1100 incorporates specialized anti-aliasing circuitry which typically attenuates aliasing products by  $\geq$  60dB; however, extremely sensitive systems may still have to take precautions to avoid aliasing errors. For more information, see the LTC1051/LTC1053 data sheet.

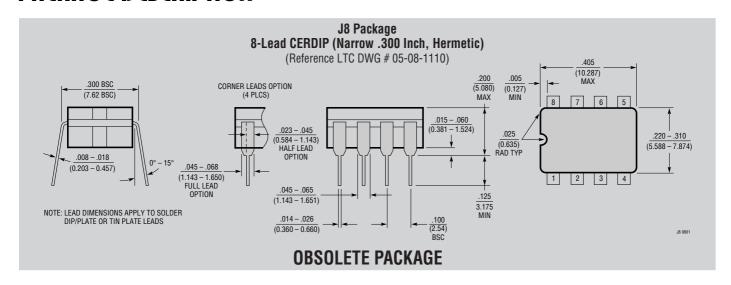
#### **Single Supply Operation**

The LTC1100 will operate on a single 5V supply, and the common mode range of the internal op amps includes ground; single supply operation is limited only by the output swing of the op amps. The internal inverting amplifier has a negative saturation limit of 5mV typically, setting the minimum common mode limit at 5mV/1.01 (or 1.1 for gain of 10). The inputs can be biased above ground, as shown in Figure 3. Low cost biasing components can be used since any errors appear as a common mode term and are rejected.

The minimum differential input voltage is limited by the swing of the output op amp. Lightly loaded, it will swing down to 5mV, allowing differential input voltages as low as  $50\mu V$  ( $450\mu V$  for gain of 10). Single supply operation limits the LTC1100 to positive differential inputs only; negative inputs will give a saturated zero output.

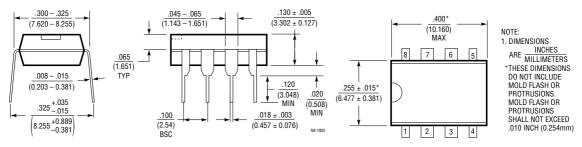


## PACKAGE DESCRIPTION



#### N8 Package 8-Lead PDIP (Narrow .300 Inch)

(Reference LTC DWG # 05-08-1510)



#### **SW Package** 16-Lead Plastic Small Outline (Wide .300 Inch)

(Reference LTC DWG # 05-08-1620)

